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(54) **Transparent embedment of data in a video signal**

(57) A method and apparatus for transparently embedding data into a video signal modulates a particular carrier frequency with a randomized pattern representing the data. The randomized pattern may be a combination of the data with the output from a pseudo-random binary sequence generator, or the data may be converted first into a bi-orthogonal pattern selected from among a plurality of such patterns and then randomized. The resulting modulated carrier frequency is then amplitude modulated based upon a human visual model so that the modulated carrier frequency amplitude is greater in

areas of high image complexity and lower in areas of low image complexity. The subliminal data represented by the modulated carrier frequency is added to the video signal. At a decoder the data is recovered using appropriate correlation techniques by first filtering the video signal to increase the subliminal data component relative to the image component, then demodulating the data component to produce the data pattern, and then correlating the data pattern with the possible data patterns to identify the particular pattern and associated data word.

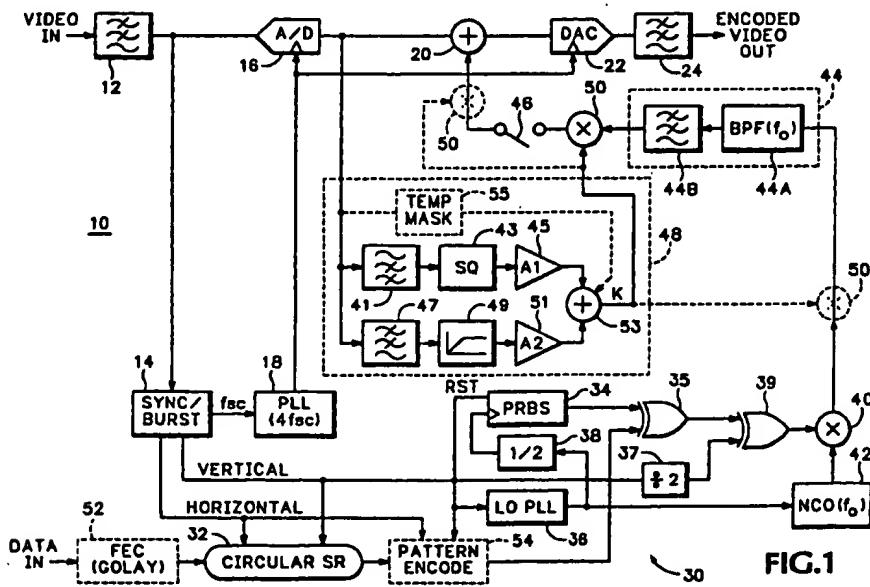


FIG.1

Description**CROSS-REFERENCE TO RELATED APPLICATIONS**

Not applicable

STATEMENT REGARDING FEDERALLY SPONSORED RESEARCH OR DEVELOPMENT

Not applicable

BACKGROUND OF THE INVENTION

The present invention relates to data transmission, and more particularly to a method and apparatus for transparently embedding data within a video signal in order to indicate authentication, program ownership and/or reception-verification of the video signal.

Television signals are usually copyrighted or otherwise proprietary to the originator and, in the case of television network distribution, are distributed to affiliate stations for re-broadcast. Unauthorized re-broadcast is difficult to detect since it is often difficult to determine the creator or originator of the material from the material itself. This is particularly true for short sequences or when the video image has been cropped or modified to mask any identification logos that may have been explicitly burned into the video image.

Another common problem with television and video distribution, which is becoming more severe, is the synchronization of an aural or audio component that is to be distributed with the video signal. Modern digital signal processing techniques requiring large buffering, such as MPEG compression, add latency to the video distribution and, since the audio may be distributed or processed separately, an error in audio to video "lip-sync" or in sound to action often occurs. Sometimes this latency is variable, requiring continual re-synchronization. In addition to audio to video synchronization problems, the audio signal may be distorted or missing entirely. By embedding, for example, the audio envelope as data within the video signal, it is also possible to detect and compare the received audio with the original which was coded and embedded as data within the video signal as a quality measure as well as video to audio delay.

Another problem is with authentication of video signals representing visual images, possibly created by a computer, where it is desirable to detect a forgery or imposter signal. In this case the video signal may be replicated to such a degree that it is difficult to detect it from the authentic video sequence. Also it is sometimes desirable to detect the beginning and/or ending of a particular sub-sequence of a video signal, such as a motion test sequence for in-service video quality assessment of video processed by MPEG compression. It is possible to devise a remote receiver to capture a special test sequence segment of a distributed video signal and compare that segment to an undistorted, stored version to

assess the quality of the video. Although it is possible to uniquely and reliably detect when the sequence occurs so that it, and only it, is captured, this generally requires the advance processing of a signature vector, transmission of that vector through a separate channel and a correlation means at the receiver. By transparently embedding data into the video a predetermined start and end code may be detected by the receiver without any preloading of signature vectors and signature vector preprocessing.

Published Canadian Patent Application No. 2,174,413 (A1) by Geoffrey B. Rhoads entitled "Identification/Authentication Coding Method and Apparatus" discusses techniques for providing authentication of image signals. In Rhoads an imperceptible N-bit identification code is embedded throughout an image with a small noise pattern in a coded fashion. In particular bits of a binary identification code are referenced sequentially to add up to N independent, noise-like patterns to the original image signal. The detection of these patterns is done by N sequential correlations with stored replicas of each pattern. This may also be done simultaneously by N correlators, as is well described in the public domain as a "correlation receiver." Rhoads further discloses adding or subtracting exactly N independent, noise-like images to improve detection and/or encoded image quality. This later modification, referred to by Rhoads as "true polarity", is well described in the public domain as "bi-orthogonal signaling" for a correlation receiver allowing 2^N composite symbols (patterns) to be created by adding or subtracting N bi-orthogonal symbols (patterns). A disadvantage of both of Rhoads' methods is that each of the N noise-like patterns, which are added or subtracted to form one of the 2^N possible composite patterns, needs to be properly scaled and designed to minimize image degradation by the addition of the composite pattern to the image. A further disadvantage is that the original unencoded image, as well as the N patterns, need to be stored in the receiver so that the unencoded image may be subtracted from the encoded image for detection.

U.S. Patent No. 4,969,041 issued November 6, 1990 to William J. O'Grady and Robert J. Dubner entitled "Embedment of Data in a Video Signal" discloses adding one of a plurality of low-level waveforms to the video signal, with the level of the low level waveform being below the noise level of the video signal, for authentication purposes or for transmitting information. Each low-level waveform corresponds to a particular data word being embedded. At the receiving end the video signal is correlated with an identical set of low-level waveforms to produce a set of correlation coefficients -- the highest correlation coefficient indicating the presence of a particular one of the low-level waveforms which is then converted into the corresponding data word. The Rhoads composite pattern consists of the summation of up to N independent patterns, which would be the same as O'Grady/Dubner when N=1 since

only one pattern is sent at a time. For an N-bit data word O'Grady/Dubner implies the need for storage of 2^N patterns rather than the N patterns of Rhoads. But since 2^N patterns may be generated by summing N bi-orthogonal sub-patterns, the effect is the same and only N patterns need to be stored in a manner identical to Rhoads. In both of these patents the symbol rate is one pattern per picture or field.

Some other drawbacks of O'Grady/Dubner are that it limits the degree to which a pattern or sequence of patterns can be hidden by not fully exploiting the limitations of the human psychovisual process, such as spatial masking; it does not fully exploit the available signal to noise ratio since the correlation output is unipolar; and it does not include coding of the data represented by the embedded symbols so as to provide temporal redundancy for error correction over missing video frames.

What is desired is a method of embedding an unobtrusive data pattern into the video signal that is useful in addressing the above-noted problems.

BRIEF SUMMARY OF THE INVENTION

Accordingly the present invention provides a method and apparatus for transparently embedding data into a video signal by modulating a particular carrier frequency with a randomized pattern representing the data. The randomized pattern may be a combination of the data with the output from a pseudo-random binary sequence generator, or the data may be converted first into a bi-orthogonal pattern selected from among a plurality of such patterns and then randomized. The resulting modulated carrier frequency is then amplitude modulated based upon a human visual model so that the modulated carrier frequency amplitude is greater in areas of high image complexity and lower in areas of low image complexity. The subliminal data represented by the modulated carrier frequency is added to the video signal. At a decoder the data is recovered using appropriate correlation techniques by first filtering the video signal to increase the subliminal data component relative to the image component, then demodulating the data component to produce the data pattern, and then correlating the data pattern with the possible data patterns to identify the particular pattern and associated data word.

The objects, advantages and other novel features of the present invention are apparent from the following detailed description when read in conjunction with the appended claims and attached drawing.

BRIEF DESCRIPTION OF THE SEVERAL VIEWS OF THE DRAWING

Fig. 1 is a block diagram view of an encoder for transparently embedding data in a video signal according to the present invention.

Fig. 2 is a block diagram view of a feedback model for determining parameter values for a human visual

model according to the present invention.

Fig. 3 is a block diagram view of a decoder for extracting embedded data in the video signal according to the present invention.

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DETAILED DESCRIPTION OF THE INVENTION

The invisibility of embedded data in a video signal is achieved by several independent factors acting together. The first factor is the principle of spectral spreading from Direct-Sequence Spread Spectrum (DSSS) modulation of a carrier frequency. The low-bandwidth/low-rate data patterns or symbols are modulated by a higher rate Pseudo Random Binary Sequence (PRBS), spreading the power spectrum around the carrier frequency so that the time-domain waveform appears noise-like or lacks the spatial coherence that the eye/brain is good at seeing even at low levels. Also by careful selection of the carrier frequency and phase it is possible to further reduce the visibility of the resulting spatial pattern created by the PRBS modulated symbols (patterns) on a DSSS modulated carrier.

Secondly it is also desirable to filter the embedded signal both vertically and horizontally to reduce edge visibility, further enhancing it's overall invisibility. A third factor that is significant is to exploit visual masking by varying the data signal level according to the video content.

Another factor is the selection of a Field Alternate Carrier (FAC) frequency and phase which becomes the DSSS modulated carrier. The invisibility of this carrier is a function of it's frequency, but it's phase from one field to the next is very important. A first-order cancellation occurs by repeating the entire data pattern on the even field (field 1) again on the odd field (field 2), but with an inverted carrier phase, causing an interlaced line pair cancellation. By repeating the data in each field, an even-line field comb filter may be used in a decoder to cancel the field to field correlated video and recover the data pattern with a 3 dB advantage at a cost of a 50% data-rate penalty. For progressive scan video systems the Field Alternate Carrier frequency and data repeat may be replaced by a Line Alternate Carrier frequency and data repeat. This gives the same or better visual cancellation and has the same data rate as the interlaced scan system.

With video signals the decoder is greatly simplified since carrier, code and data lock are simply derived from the abundant timing information in sync and burst. In addition to the rejection afforded by the comb filter, a band-pass filter is used to suppress the large carrier components of the program video signal. After the PRBS correlation, or de-spreading, various detection methods are possible. An integrate and dump matched filter driving a software, bi-orthogonal M-ary scheme may be used. Two options with and without channel coding also may be used. Numerous alternatives include combining convolution coding with Walsh code assignments (Unger-

oech TCM) and concatenated RS and convolution.

Channel coding is best done in a holographic, two-dimensional (2D) sense by shuffling or the like for robustness against video motion of pixel segments passed through the comb filter. Such 2D holographic shuffling of data may be extended to three-dimensional (3D) shuffling because video has a time sense unlike still pictures. This may be done by repeating some of the data over several frames, which could be valuable for video sequence identification. The data that identifies the video sequence may be given a lower energy, i.e., fewer pixels, than the data that identifies the particular frame within the video sequence, since the frames come only once while the sequence has many frames in it. Data security through data encryption may always be applied separately from the channel coding, allowing the decoder design to be public or even standardized.

Referring now to Fig. 1 an encoder 10 is shown having an analog video signal which is input to an anti-aliasing, lowpass filter 12, the output of which is input to a sync/burst separator 14 to generate necessary timing signals and to an analog to digital converter (ADC) 16 to produce a digital video signal. A sample clock phase locked loop (PLL) 18 in response to one of the timing signals (subcarrier frequency fsc) from the sync/burst generator 14 provides a sample clock at four times subcarrier (4fsc). Where the input signal is a digital video signal, the timing signals are extracted from the digital video signal and the lowpass filter 12 and ADC 16 are not needed. The digital video signal is input to an adder 20 where a subliminal data signal is added to the digital video signal. The resulting digital video signal is input to a digital to analog converter (DAC) 22 to produce a modulated analog video signal which in turn is processed by an output lowpass reconstruction filter 24 to produce a final encoded analog video signal. For digital distribution the DAC 22 and filter 24 may be eliminated.

A modulator circuit 30 generates the subliminal data signal for input to the adder 20. The modulator circuit 30 has as inputs the vertical and horizontal timing signals from the sync/burst separator 14 and the desired subliminal data to be added to the digital video signal. The subliminal data is input to a circular shift register 32 which has as inputs the horizontal and vertical sync signals. The subliminal data from the circular shift register 32 may be input to an exclusive OR gate 34 where it is combined with the output from a pseudo random binary sequence (PRBS) generator 34. The PRBS generator 34 is reset by the vertical sync signal and clocked by a clock signal derived via a local PLL 36 and a divider 38. The output from the exclusive OR gate 35 is a series of +/-1's which is then input to a mixer 40. A numerically controlled oscillator (NCO) 42 provides a modulator frequency fo that also is input to the mixer 40. The output from the mixer 40 is a modulated frequency signal which is the DSSS signal as modulated by the output from the PRBS generator 34 and the data from the circular shift register 32.

To provide the Field Alternate Carrier frequency an inversion of the phase of either of the inputs to the modulator 40 is performed using a divide-by-two circuit 37 having the vertical interval timing signal as an input followed by another exclusive OR gate 39. The exclusive OR gate 39 has as shown the other input coupled to the output of the first exclusive OR gate 35, the output being coupled to the modulator 40. Alternatively the exclusive OR gate 39 may be inserted between the NCO 42 and

10 the modulator 40, between the circular shift register 32 and the first exclusive OR gate 35, or between the PRBS generator 34 and the first exclusive OR gate. For progressive scan systems the input to the divide-by-two circuit 37 would be the horizontal interval timing signal so 15 that the Line Alternate Carrier frequency is implemented. In either case alternate lines in each complete frame of the video image sequence have the same data, but inverted in phase, so that subtraction of alternate lines by a comb filter in a decoder enhances the subliminal 20 data signal.

The modulated frequency is input to two-dimensional, horizontal and vertical spatial frequency, softening filters 44. The horizontal filter 44A is centered about the modulator frequency fo to mitigate any fast edges in 25 the modulated frequency signal, and the vertical filter 44B mitigates fast vertical edges (line-to-line transitions). The modulated frequency signal is the subliminal data signal that is input to the adder 20 via a blanking gate 46 so that the addition occurs only during the active 30 video portion of the digital video signal.

A human vision model (HVM) circuit 48 has as inputs the digital video signal at the input to the adder 20. One practical implementation of visual masking based upon the human visual model is shown. A first aspect of 35 the human visual model is noise tolerance in areas of high spatial detail, and this is controlled by a two-dimensional highpass filter 41, such as a Sobel filter, followed by an envelope squarer 43 and a first control amplifier 45. A second aspect is sensitivity to noise at different 40 local contrasts, and this is controlled by a two-dimensional lowpass filter 47 followed by a transfer offset circuit 49 and a second control amplifier 51. The parameters A1 and A2 represented by the first and second control amplifiers 45, 51 are adjusted for a desired minimum 45 data pattern visibility over a range of picture material. A third aspect may be that of temporal masking 55 where noise on screen cuts or motion is temporarily less visible. The resulting outputs from these paths, describing 50 the complexity of the image rendered by the digital video signal, are summed by a summer 53 to produce a gain signal to control a multiplier 50 to which the modulated frequency signal is input prior to the adder 20. Thus for portions of the digital video signal where the image is unchanging in motion and/or has low spatial detail and contrast or color/shading variations, the gain is relatively low; and where the image is complex due to motion and/or high spatial detail, the gain is relatively high. In fact 55 the amplitude of the subliminal data signal does not

need to be less than the noise floor in complex regions of the video digital signal to retain an acceptable level of invisibility.

A feedback model 60 for determining the parameter values **A1**, **A2** represented by the first and second control amplifiers 45, 51 and the transfer curve of the transfer-offset circuit 49 is shown in Fig. 1. A desired or acceptable distortion value is pre-set and dynamically maintained by feedback based on the human visual model. An image sequence is stored in a video buffer 62 and repeatedly output in order to compare the output from the human visual model 64 prior to the encoder 10 with the image sequence derived from the output of the human visual model 66 following the data encoder. The comparison between the human visual model outputs is performed by means of a means absolute difference or mean square error algorithm 68. The resulting difference or error is input to a comparator 69 to which also is input a distortion level which is adjustable to the desired level of invisibility. The parameters of the first and second control amplifiers 45, 51 and the transfer curve of the transfer-offset circuit 49 are adjusted to maintain comparator 69 output high. This is repeated over a range of exemplary image sequences. This then produces an acceptable estimate of these parameters which may be fixed for all video to be encoded. The human visual model is only as elaborate as necessary to match the dominant elements of the human visual system as it relates to the desired level of masking.

This feedback model may also be used without the masking modulator 48 in an iterative manner to directly determine the pattern level gain signal for input to the multiplier 50. In other words a vector or array of gain signal values, maximized for each pixel, may be determined based on the human visual model feedback, optimizing the root mean square (RMS) value of the embedded pseudo-noise pattern while maintaining the desired level of distortion.

Forward error correction may be added to the modulator 30 by inputting the subliminal data to a forward error correction (FEC) circuit 52 prior to input into the circular shift register 32. This will provide error correction for bit or byte errors when the data from the circular shift register 32 is input directly to exclusive OR gate 35. The FEC circuit 52 may use a bit-error correction code, such as Golay codes, as the forward error correction algorithm to improve the data integrity for bit errors, or a more sophisticated channel coding such as concatenated convolution and Reed-Solomon (RS) code may be used, as is well known in the art. These coding techniques combined with bit shuffling provide error correction across frames so that entire frames may be lost without any loss in data. For example FEC 52 may be designed to provide RS word-error correction where each frame contains a coded binary word of a fixed length of m data bits, allowing missing words (frames) to be completely recovered from the redundancy in adjacent words (frames). In this case the output from the

circular shift register 32 may be input to an orthogonal pattern selector 54, such as a Walsh encoder, to produce the subliminal data coded to select one of a set of orthogonal patterns input to the exclusive OR gate 35.

5 Walsh codes are one-dimensional orthogonal sequences used in the well-known IS-95 Code Division Multiple Access wireless telephone standard. Also Walsh sequences are the basis for a particularly well-known set of two-dimensional orthogonal patterns called Hadamard patterns. Orthogonal patterns are those that, when correlated with all other patterns in the set, produce a positive correlation coefficient for only one of the patterns, with all of the other patterns producing a zero correlation coefficient. Where each pattern represents 10 m bits, a total of 2^m patterns are required and the data from the circular shift register 32 acts as an address to the pattern selector 54. By using bi-orthogonal patterns 15 only 2^{m-1} patterns or sequences are needed for m bits. These patterns may be input directly to the mixer 20 40 without the need of exclusive OR gate 35, but the bi-orthogonal patterns or Walsh sequences may not all appear visually random, which would make them more visible when added to the program video signal. However when this pattern is exclusive OR'd with the PRBS generator 25 output, the resulting light pattern at the output of the modulator 40 becomes much more random in appearance and is difficult to distinguish from noise. The resulting modulation is a randomized bi-phase modulation spreading the spectral energy around the carrier. 30 Circular shift register 32 allows the data or data pattern to be repeated on interlaced fields so that the visual line pair cancellation and field comb data separator in the decoder may be used, as described above.

A decoder 70 is shown in Fig. 3. The input section 35 72 is the same as for the encoder 10, having a lowpass anti-aliasing filter 74, a sync/burst separator 76 to generate the timing signals, a PLL 78 to generate the sample clock and an analog to digital converter 80 to produce the digital video signal. The decoder 70 also includes a Local Oscillator (LO) section 82 which includes a PRBS generator 84 with PLL 86 and divider 88 and a numerically controlled oscillator (NCO) 90 that produces the demodulation carrier frequency f_0 . The outputs from the PRBS generator 84 and the NCO 90 are input to a mixer 91 to produce a pseudo random demodulated LO signal (BPSK -- Bi-Phase Shift Keyed) for a down-converter mixer 102.

50 The digital video signal is input to preprocessing filters 92 that include a bandpass filter 94 centered at the modulator frequency and a high pass field-comb filter 96 including a field delay 98 and a subtractor 100. These filters remove much of the program video signal, enhancing the detection of the subliminal embedded data signal. The field-comb filter 96 is particularly effective 55 since the embedded data pattern is repeated with inverse polarity on adjacent fields, doubling its level at the filter output while the program video signal that is correlated on adjacent fields is removed. The filtered digital

video signal is input to the down-converter mixer 102 and mixed together with the pseudo random modulated LO frequency signal. The resulting baseband digital signal is input to a correlation/matched filter circuit 104.

The correlation/matched filter circuit 104 has as inputs the baseband digital signal and the horizontal and vertical interval timing signals. The timing signals are input to an orthogonal pattern generator 106, such as a Walsh generator, that is comparable to that in the encoder 10. The timing signals are also input to a matched filter 108 to which also is input the baseband digital signal. The matched filter 108 includes an integrate and dump circuit 110 followed by a sampler 112. If the embedded data is only as bits encoded with an FEC channel code, as discussed previously, matched filter 108 output may be converted to binary data by slicer 124 and input directly to FEC decoder 126 to output the decoded and error corrected subliminal data. In the case where the data is coded as bi-orthogonal patterns, the output from the sampler 112 is input to a bi-orthogonal correlation receiver 114 that includes a correlation mixer 116 for each of the patterns produced by the pattern generator 106 so that the sampler output is correlated with each pattern. The outputs from the correlation mixers 116 are integrated by respective integrators 118, the outputs of which are input to a maximum magnitude selector 120. Due to the orthogonality of the patterns only one of the outputs from the integrators 118 has a significant magnitude that is detected by the magnitude and magnitude polarity selector 120 as one of the predetermined m bit patterns or data words assigned to each pattern. These m data words may have been FEC coded as described previously. In this case the error corrector 126 is designed to correct word errors based upon the redundancy in other decoded data words in order to correct data errors from missing video frames or from heavily distorted embedded data patterns. The magnitude selector 120 detects not only the greatest magnitude, but the polarity of that magnitude to provide an extra bit. In other words eight patterns represent three bits plus one bit for the eight inverted patterns, providing $m=4$ bits.

The subliminal data channel provided by the present invention may be used for stamping an identifying mark on a video sequence even if some of the video frames have been lost or portions of video frames have been cropped or distorted through lossy compression/decompression processing. The identifying mark may represent an audio signal or at least a signature of the audio into the active video to allow a remote video receiver to compare the audio signature with the received audio to automatically advance or delay the relative audio to video timing to maintain sound to video synchronization as encoded at the source. The subliminal data channel may represent modern cryptographic digital signature authentication in the video. It may be used to identify a motion test sequence, or portion thereof, to be captured for video quality assessment, such as

discussed in co-pending U.S. Patent Application No. 08/605,241 filed February 12, 1996 by Bozidar Janko and David K. Fibush entitled "Programmable Instrument for Automatic Measurement of Compressed Video

5 Quality". The subliminal data channel also may be used to analyze timeliness of service, i.e., did the video get aired at the desired time in its entirety?

Thus the present invention provides for transparently encoding data into a video image by modulating a 10 particular carrier frequency, providing interlaced field visual masking with a unique one of $2^{(m-1)}$ bi-orthogonal patterns, the pattern representing m bits, randomizing the pattern with a PRBS sequence to reduce visibility, amplitude modulating the random pattern modulated 15 carrier to further maximize embedded signal amplitude and minimize visibility based on a human visual model exploiting visual masking, and adding the resulting pattern modulated frequency signal to the video such that the data is recoverable using appropriate correlation 20 techniques.

Claims

25 1. An apparatus for transparently encoding data within a video signal comprising:

means for generating from the data to be encoded and video timing signals associated with the video signal a randomized binary pattern; 30 means for modulating the randomized binary pattern onto a carrier frequency to produce a subliminal data signal; and means for combining the subliminal data signal with the video signal to produce an output video signal having the data transparently encoded within the video signal.

35 2. The apparatus as recited in claim 1 further comprising means for alternating the phase of the subliminal data signal on alternate lines of each frame of the video signal such that each pair of contiguous lines has the same data but inverted in phase.

40 45 3. The apparatus as recited in claim 2 wherein the alternating means comprises:

means for dividing a specific one of the video timing signals by two to produce a phase inversion control signal; and 50 means for inverting the phase of the subliminal data signal on alternate lines of each frame in response to the phase inversion signal.

55 4. The apparatus as recited in claim 3 wherein for an interlaced version of the video signal the specific one of the video timing signals is a field rate signal and for a progressive scan version of the video sig-

nal the specific one of the video timing signals is a line rate signal.

5. The apparatus as recited in claim 1 further comprising means for amplitude modulating the subliminal data signal prior to the combining means with a gain signal based upon a human visual model.

6. The apparatus as recited in claims 1, 2 or 5 further comprising means for softening edges in the subliminal data signal prior to the combining means.

7. A method of transparently encoding data within a video signal comprising the steps of:

generating from the data to be encoded and video timing signals associated with the video signal a randomized binary pattern; modulating the randomized binary pattern onto a carrier frequency to produce a subliminal data signal; and combining the subliminal data signal with the video signal to produce an output video signal having the data transparently encoded within the video signal.

8. The method as recited in claim 7 further comprising the step of alternating the phase of the subliminal data signal on alternate lines of each frame of the video signal such that each pair of contiguous lines has the same data but inverted in phase.

9. The method as recited in claim 8 wherein the alternating step comprises the steps of:

dividing a specific one of the video timing signals by two to produce a phase inversion control signal; and inverting the phase of the subliminal data signal on alternate lines of each frame in response to the phase inversion signal.

10. The method as recited in claim 9 wherein for an interlaced version of the video signal the specific one of the video timing signals is a field rate signal and for a progressive scan version of the video signal the specific one of the video timing signals is a line rate signal.

11. The method as recited in claim 7 further comprising the step of amplitude modulating the subliminal data signal prior to the combining step with a gain signal based upon a human visual model.

12. The method as recited in claims 7, 8 or 11 further comprising the step of softening edges in the subliminal data signal prior to the combining step.

13. The method as recited in claim 7 further comprising the step of shuffling the data two-dimensionally to provide channel coding.

5 14. The method as recited in claim 7 further comprising the step of shuffling the data three-dimensionally to provide channel coding.

10 15. An apparatus for transparently embedding data within a video signal comprising:

a pattern generator having as inputs timing signals associated with the video signal and the data to be embedded and producing as an output a randomized pattern signal; a modulator having as inputs the randomized pattern signal and a carrier frequency synchronized with the randomized pattern signal and producing as an output a subliminal data signal; and a combiner having as inputs the subliminal data signal and the video signal and producing as an output a transparently embedded data video signal.

15 16. The apparatus as recited in claim 15 further comprising a phase alternator circuit having as inputs a specific one of the video timing signals and a component of the subliminal data signal and having as an output a signal that produces inversion of the phase of the subliminal data signal on alternate lines of each frame of the video signal such that each pair of contiguous lines has the same data but inverted in phase.

20 25 30 35 40 45 50 55 60 65 70 75 80 85 90 95 100 105 110 115 120 125 130 135 140 145 150 155 160 165 170 175 180 185 190 195 200 205 210 215 220 225 230 235 240 245 250 255 260 265 270 275 280 285 290 295 300 305 310 315 320 325 330 335 340 345 350 355 360 365 370 375 380 385 390 395 400 405 410 415 420 425 430 435 440 445 450 455 460 465 470 475 480 485 490 495 500 505 510 515 520 525 530 535 540 545 550 555 560 565 570 575 580 585 590 595 600 605 610 615 620 625 630 635 640 645 650 655 660 665 670 675 680 685 690 695 700 705 710 715 720 725 730 735 740 745 750 755 760 765 770 775 780 785 790 795 800 805 810 815 820 825 830 835 840 845 850 855 860 865 870 875 880 885 890 895 900 905 910 915 920 925 930 935 940 945 950 955 960 965 970 975 980 985 990 995 1000 1005 1010 1015 1020 1025 1030 1035 1040 1045 1050 1055 1060 1065 1070 1075 1080 1085 1090 1095 1100 1105 1110 1115 1120 1125 1130 1135 1140 1145 1150 1155 1160 1165 1170 1175 1180 1185 1190 1195 1200 1205 1210 1215 1220 1225 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2230 2235 2240 2245 2250 2255 2260 2265 2270 2275 2280 2285 2290 2295 2300 2305 2310 2315 2320 2325 2330 2335 2340 2345 2350 2355 2360 2365 2370 2375 2380 2385 2390 2395 2400 2405 2410 2415 2420 2425 2430 2435 2440 2445 2450 2455 2460 2465 2470 2475 2480 2485 2490 2495 2500 2505 2510 2515 2520 2525 2530 2535 2540 2545 2550 2555 2560 2565 2570 2575 2580 2585 2590 2595 2600 2605 2610 2615 2620 2625 2630 2635 2640 2645 2650 2655 2660 2665 2670 2675 2680 2685 2690 2695 2700 2705 2710 2715 2720 2725 2730 2735 2740 2745 2750 2755 2760 2765 2770 2775 2780 2785 2790 2795 2800 2805 2810 2815 2820 2825 2830 2835 2840 2845 2850 2855 2860 2865 2870 2875 2880 2885 2890 2895 2900 2905 2910 2915 2920 2925 2930 2935 2940 2945 2950 2955 2960 2965 2970 2975 2980 2985 2990 2995 3000 3005 3010 3015 3020 3025 3030 3035 3040 3045 3050 3055 3060 3065 3070 3075 3080 3085 3090 3095 3100 3105 3110 3115 3120 3125 3130 3135 3140 3145 3150 3155 3160 3165 3170 3175 3180 3185 3190 3195 3200 3205 3210 3215 3220 3225 3230 3235 3240 3245 3250 3255 3260 3265 3270 3275 3280 3285 3290 3295 3300 3305 3310 3315 3320 3325 3330 3335 3340 3345 3350 3355 3360 3365 3370 3375 3380 3385 3390 3395 3400 3405 3410 3415 3420 3425 3430 3435 3440 3445 3450 3455 3460 3465 3470 3475 3480 3485 3490 3495 3500 3505 3510 3515 3520 3525 3530 3535 3540 3545 3550 3555 3560 3565 3570 3575 3580 3585 3590 3595 3600 3605 3610 3615 3620 3625 3630 3635 3640 3645 3650 3655 3660 3665 3670 3675 3680 3685 3690 3695 3700 3705 3710 3715 3720 3725 3730 3735 3740 3745 3750 3755 3760 3765 3770 3775 3780 3785 3790 3795 3800 3805 3810 3815 3820 3825 3830 3835 3840 3845 3850 3855 3860 3865 3870 3875 3880 3885 3890 3895 3900 3905 3910 3915 3920 3925 3930 3935 3940 3945 3950 3955 3960 3965 3970 3975 3980 3985 3990 3995 4000 4005 4010 4015 4020 4025 4030 4035 4040 4045 4050 4055 4060 4065 4070 4075 4080 4085 4090 4095 4100 4105 4110 4115 4120 4125 4130 4135 4140 4145 4150 4155 4160 4165 4170 4175 4180 4185 4190 4195 4200 4205 4210 4215 4220 4225 4230 4235 4240 4245 4250 4255 4260 4265 4270 4275 4280 4285 4290 4295 4300 4305 4310 4315 4320 4325 4330 4335 4340 4345 4350 4355 4360 4365 4370 4375 4380 4385 4390 4395 4400 4405 4410 4415 4420 4425 4430 4435 4440 4445 4450 4455 4460 4465 4470 4475 4480 4485 4490 4495 4500 4505 4510 4515 4520 4525 4530 4535 4540 4545 4550 4555 4560 4565 4570 4575 4580 4585 4590 4595 4600 4605 4610 4615 4620 4625 4630 4635 4640 4645 4650 4655 4660 4665 4670 4675 4680 4685 4690 4695 4700 4705 4710 4715 4720 4725 4730 4735 4740 4745 4750 4755 4760 4765 4770 4775 4780 4785 4790 4795 4800 4805 4810 4815 4820 4825 4830 4835 4840 4845 4850 4855 4860 4865 4870 4875 4880 4885 4890 4895 4900 4905 4910 4915 4920 4925 4930 4935 4940 4945 4950 4955 4960 4965 4970 4975 4980 4985 4990 4995 5000 5005 5010 5015 5020 5025 5030 5035 5040 5045 5050 5055 5060 5065 5070 5075 5080 5085 5090 5095 5100 5105 5110 5115 5120 5125 5130 5135 5140 5145 5150 5155 5160 5165 5170 5175 5180 5185 5190 5195 5200 5205 5210 5215 5220 5225 5230 5235 5240 5245 5250 5255 5260 5265 5270 5275 5280 5285 5290 5295 5300 5305 5310 5315 5320 5325 5330 5335 5340 5345 5350 5355 5360 5365 5370 5375 5380 5385 5390 5395 5400 5405 5410 5415 5420 5425 5430 5435 5440 5445 5450 5455 5460 5465 5470 5475 5480 5485 5490 5495 5500 5505 5510 5515 5520 5525 5530 5535 5540 5545 5550 5555 5560 5565 5570 5575 5580 5585 5590 5595 5600 5605 5610 5615 5620 5625 5630 5635 5640 5645 5650 5655 5660 5665 5670 5675 5680 5685 5690 5695 5700 5705 5710 5715 5720 5725 5730 5735 5740 5745 5750 5755 5760 5765 5770 5775 5780 5785 5790 5795 5800 5805 5810 5815 5820 5825 5830 5835 5840 5845 5850 5855 5860 5865 5870 5875 5880 5885 5890 5895 5900 5905 5910 5915 5920 5925 5930 5935 5940 5945 5950 5955 5960 5965 5970 5975 5980 5985 5990 5995 6000 6005 6010 6015 6020 6025 6030 6035 6040 6045 6050 6055 6060 6065 6070 6075 6080 6085 6090 6095 6100 6105 6110 6115 6120 6125 6130 6135 6140 6145 6150 6155 6160 6165 6170 6175 6180 6185 6190 6195 6200 6205 6210 6215 6220 6225 6230 6235 6240 6245 6250 6255 6260 6265 6270 6275 6280 6285 6290 6295 6300 6305 6310 6315 6320 6325 6330 6335 6340 6345 6350 6355 6360 6365 6370 6375 6380 6385 6390 6395 6400 6405 6410 6415 6420 6425 6430 6435 6440 6445 6450 6455 6460 6465 6470 6475 6480 6485 6490 6495 6500 6505 6510 6515 6520 6525 6530 6535 6540 6545 6550 6555 6560 6565 6570 6575 6580 6585 6590 6595 6600 6605 6610 6615 6620 6625 6630 6635 6640 6645 6650 6655 6660 6665 6670 6675 6680 6685 6690 6695 6700 6705 6710 6715 6720 6725 6730 6735 6740 6745 6750 6755 6760 6765 6770 6775 6780 6785 6790 6795 6800 6805 6810 6815 6820 6825 6830 6835 6840 6845 6850 6855 6860 6865 6870 6875 6880 6885 6890 6895 6900 6905 6910 6915 6920 6925 6930 6935 6940 6945 6950 6955 6960 6965 6970 6975 6980 6985 6990 6995 7000 7005 7010 7015 7020 7025 7030 7035 7040 7045 7050 7055 7060 7065 7070 7075 7080 7085 7090 7095 7100 7105 7110 7115 7120 7125 7130 7135 7140 7145 7150 7155 7160 7165 7170 7175 7180 7185 7190 7195 7200 7205 7210 7215 7220 7225 7230 7235 7240 7245 7250 7255 7260 7265 7270 7275 7280 7285 7290 7295 7300 7305 7310 7315 7320 7325 7330 7335 7340 7345 7350 7355 7360 7365 7370 7375 7380 7385 7390 7395 7400 7405 7410 7415 7420 7425 7430 7435 7440 7445 7450 7455 7460 7465 7470 7475 7480 7485 7490 7495 7500 7505 7510 7515 7520 7525 7530 7535 7540 7545 7550 7555 7560 7565 7570 7575 7580 7585 7590 7595 7600 7605 7610 7615 7620 7625 7630 7635 7640 7645 7650 7655 7660 7665 7670 7675 7680 7685 7690 7695 7700 7705 7710 7715 7720 7725 7730 7735 7740 7745 7750 7755 7760 7765 7770 7775 7780 7785 7790 7795 7800 7805 7810 7815 7820 7825 7830 7835 7840 7845 7850 7855 7860 7865 7870 7875 7880 7885 7890 7895 7900 7905 7910 7915 7920 7925 7930 7935 7940 7945 7950 7955 7960 7965 7970 7975 7980 7985 7990 7995 8000 8005 8010 8015 8020 8025 8030 8035 8040 8045 8050 8055 8060 8065 8070 8075 8080 8085 8090 8095 8100 8105 8110 8115 8120 8125 8130 8135 8140 8145 8150 8155 8160 8165 8170 8175 8180 8185 8190 8195 8200 8205 8210 8215 8220 8225 8230 8235 8240 8245 8250 8255 8260 8265 8270 8275 8280 8285 8290 8295 8300 8305 8310 8315 8320 8325 8330 8335 8340 8345 8350 8355 8360 8365 8370 8375 8380 8385 8390 8395 8400 8405 8410 8415 8420 8425 8430 8435 8440 8445 8450 8455 8460 8465 8470 8475 8480 8485 8490 8495 8500 8505 8510 8515 8520 8525 8530 8535 8540 8545 8550 8555 8560 8565 8570 8575 8580 8585 8590 8595 8600 8605 8610 8615 8620 8625 8630 8635 8640 8645 8650 8655 8660 8665 8670 8675 8680 8685 8690 8695 8700 8705 8710 8715 8720 8725 8730 8735 8740 8745 8750 8755 8760 8765 8770 8775 8780 8785 8790 8795 8800 8805 8810 8815 8820 8825 8830 8835 8840 8845 8850 8855 8860 8865 8870 8875 8880 8885 8890 8895 8900 8905 8910 8915 8920 8925 8930 8935 8940 8945 8950 8955 8960 8965 8970 8975 8980 8985 8990 8995 9000 9005 9010 9015 9020 9025 9030 9035 9040 9045 9050 9055 9060 9065 9070 9075 9080 9085 9090 9095 9100 9105 9110 9115 9120 9125 9130 9135 9140 9145 9150 9155 9160 9165 9170 9175 9180 9185 9190 9195 9200 9205 9210 9215 9220 9225 9230 9235 9240 9245 9250 9255 9260 9265 9270 9275 9280 9285 9290 9295 9300 9305 9310 9315 9320 9325 9330 9335 9340 9345 9350 9355 9360 9365 9370 9375 9380 9385 9390 9395 9400 9405 9410 9415 9420 9425 9430 9435 9440 9445 9450 9455 9460 9465 9470 9475 9480 9485 9490 9495 9500 9505 9510 9515 9520 9525 9530 9535 9540 9545 9550 9555 9560 9565 9570 9575 9580 9585 9590 9595 9600 9605 9610 9615 9620 9625 9630 9635 9640 9645 9650 9655 9660 9665 9670 9675 9680 9685 9690 9695 9700 9705 9710 9715 9720 9725 9730 9735 9740 9745 9750 9755 9760 9765 9770 9775 9780 9785 9790 9795 9800 9805 9810 9815 9820 9825 9830 9835 9840 9845 9850 9855 9860 9865 9870 9875 9880 9885 9890 9895 9900 9905 9910 9915 9920 9925 9930 9935 9940 9945 9950 9955 9960 9965 9970 9975 9980 9985 9990 9995 10000 10005 10010 10015 10020 10025 10030 10035 10040 10045 10050 10055 10060 10065 10070 10075 10080 10085 10090 10095 10100 10105 10110 10115 10120 10125 10130 10135 10140 10145 10150 10155 10160 10165 10170 10175 10180 10185 10190 10195 10200 10205 10210 10215 10220 10225 10230 10235 10240 10245 10250 10255 10260 10265 10270 10275 10280 10285 10290 10295 10300 10305 10310 10315 10320 10325 10330 10335 10340 10345 10350 10355 10360 10365 10370 10375 10380 10385 10390 10395 10400 10405 10410 10415 10420 10425 10430 10435 10440 10445 10450 10455 10460 10465 10470 10475 10480 10485 10490 10495 10500 10505 10510 10515 10520 10525 10530 10535 10540 10545 10550 10555 10560 10565 10570 10575 10

tered subliminal data signal for input to the combiner as the subliminal data signal.

20. The apparatus as recited in claims 15, 16 or 19 further comprising an amplitude modulator having as inputs the subliminal data signal and a gain control signal based upon a human visual model and producing as an output an amplitude modulated subliminal data signal as the subliminal data signal for input to the combiner. 5

21. The apparatus as recited in claim 15 wherein the pattern generator comprises:

a data pattern generator having as inputs the data to be embedded and the timing signals and producing as an output a binary pattern; 15
a pseudo random binary sequence generator having as an input one of the timing signals and producing as an output a random binary sequence; 20
a combination circuit having as inputs the binary pattern and the random binary sequence and producing as an output the randomized binary pattern.

22. The apparatus as recited in claim 21 wherein the data pattern generator comprises a circular shift register having as inputs the data and the timing signals and producing as an output the binary pattern. 25

23. The apparatus as recited in claim 22 wherein the data pattern generator further comprises a pattern encoder having the output from the circular shift register and the timing signals as inputs and producing as an output the binary pattern. 30

24. The apparatus as recited in claim 23 wherein the data pattern generator provides a plurality of bi-orthogonal binary patterns, with the one selected for output as the binary pattern being determined by the output from the circulating shift register. 35

25. The apparatus as recited in claim 23 wherein the data pattern generator comprises a Walsh encoder having as input the output from the circular shift register and producing as an output the binary pattern selected from among a plurality of bi-orthogonal patterns. 40

26. The apparatus as recited in claim 15 wherein the combination circuit comprises an exclusive OR gate having as inputs the binary pattern and the random binary sequence and producing as an output the randomized binary pattern. 45

27. The apparatus as recited in claim 19 wherein the softening filter comprises:

a horizontal bandpass filter centered on the carrier frequency having as an input the subliminal data signal and producing as an output a horizontally filtered subliminal data signal; and 50
a vertical lowpass filter having as an input the horizontally filtered subliminal data signal and producing as an output the filtered subliminal data signal.

10 28. The apparatus as recited in claim 20 further comprising:

a highpass human visual model filter having as an input the video signal and producing as an output a highpass filtered video signal representing high spatial detail; 55
a lowpass human visual model filter having as an input the video signal and producing as an output a lowpass filtered video signal representing local contrast; and
a combining circuit having as inputs the lowpass and highpass filtered video signals and producing as an output the gain signal.

25 29. The apparatus as recited in claim 28 further comprising a temporal human visual model filter having as an input the video signal and producing as an output a temporally filtered video signal, the temporally filtered video signal being applied as another input to the combining circuit so that the gain signal is a combination of the lowpass, highpass and temporally filtered video signals. 60

30 30. The apparatus as recited in claim 20 wherein the amplitude modulator is inserted between the modulator and the softening filter. 65

35 31. The apparatus as recited in claim 20 wherein the amplitude modulator is inserted between the softening filter and the combiner. 70

40 32. The apparatus as recited in claims 15, 16, or 19 further comprising a forward error correction circuit having as an input the data to be transparently encoded within the video signal and producing as an output a forward corrected data signal coupled to the input of the pattern generator. 75

45 33. An apparatus for extracting transparently embedded data from a video signal where the embedded data is a randomized binary pattern modulated onto a carrier frequency comprising:

means for demodulating the video signal about the carrier frequency with a random binary sequence modulated carrier frequency to produce a broadband video signal; and 80
means for correlating the broadband video sig-

nal with a plurality of binary patterns to recover the embedded data.

34. The apparatus as recited in claim 33 further comprising means for filtering the video signal prior to the demodulating means to enhance the embedded data with respect to the video content of the video signal.

35. The apparatus as recited in claim 34 wherein the filtering means comprises:

means centered on the carrier frequency for broadband filtering the video signal to produce a first filtered video signal; and
means for comb filtering the first filtered video signal to produce the video signal for input to the demodulating means.

36. A method of extracting transparently embedded data from a video signal where the embedded data is a randomized binary pattern modulated onto a carrier frequency comprising the steps of:

demodulating the video signal about the carrier frequency with a random binary sequence modulated carrier frequency to produce a broadband video signal; and
correlating the broadband video signal with a plurality of binary patterns to recover the embedded data.

37. The method as recited in claim 36 further comprising the step of filtering the video signal prior to the demodulating step to enhance the embedded data with respect to the video content of the video signal.

38. The method as recited in claim 37 wherein the filtering step comprises the steps of:

broadband filtering the video signal about the carrier frequency to produce a first filtered video signal; and
comb filtering the first filtered video signal to produce the video signal for input to the demodulating step.

39. An apparatus for extracting transparently embedded data from a video signal where the embedded data is a randomized binary pattern modulated onto a carrier frequency comprising:

a demodulator having as inputs a random binary sequence modulated carrier frequency and the video signal and producing as an output a broadband video signal; and
a correlation receiver having as inputs the broadband video signal and timing signals as-

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sociated with the video signal for comparing the broadband video signal with a plurality of binary patterns to recover as an output the embedded data.

40. The apparatus as recited in claim 39 further comprising a filter having as an input the video signal and producing as an output a filtered video signal coupled to the input of the demodulator, the filter enhancing the embedded data with respect to the video content of the video signal.

41. The apparatus as recited in claim 40 wherein the filter comprises:

a broadband filter centered on the carrier frequency having as an input the video signal and producing as an output a first filtered video signal; and
a field comb filter having as an input the first filtered video signal and producing as an output the filtered video signal coupled to the input of the demodulator.

42. The apparatus as recited in claim 39 wherein the demodulator comprises:

a pseudo random binary sequence generator having as an input one of the timing signals and providing as an output a random binary sequence;
a numerically controlled oscillator having as an input a control signal derived from the one timing signal and providing as an output the carrier frequency;
a first modulator having as inputs the carrier frequency and the random binary sequence and producing as an output a modulated carrier frequency; and
a second modulator having as inputs the video signal and the modulated carrier frequency and producing as an output the broadband video signal.

43. The apparatus as recited in claim 39 wherein the correlation receiver comprises:

a sampler having as an input the broadband video signal and producing as an output a sampled video signal;
a pattern buffer having as inputs the timing signals and providing as outputs in parallel a plurality of binary patterns;
a plurality of correlation circuits in parallel, each having as inputs the sampled video signal and a specific one of the plurality of binary patterns and producing as an output a specific correlation coefficient; and

a magnitude comparator having as inputs the correlation coefficients from each correlation circuit and producing as an output a binary data word representing the sign and value of the greatest of the correlation coefficients, the binary data word being the embedded data. 5

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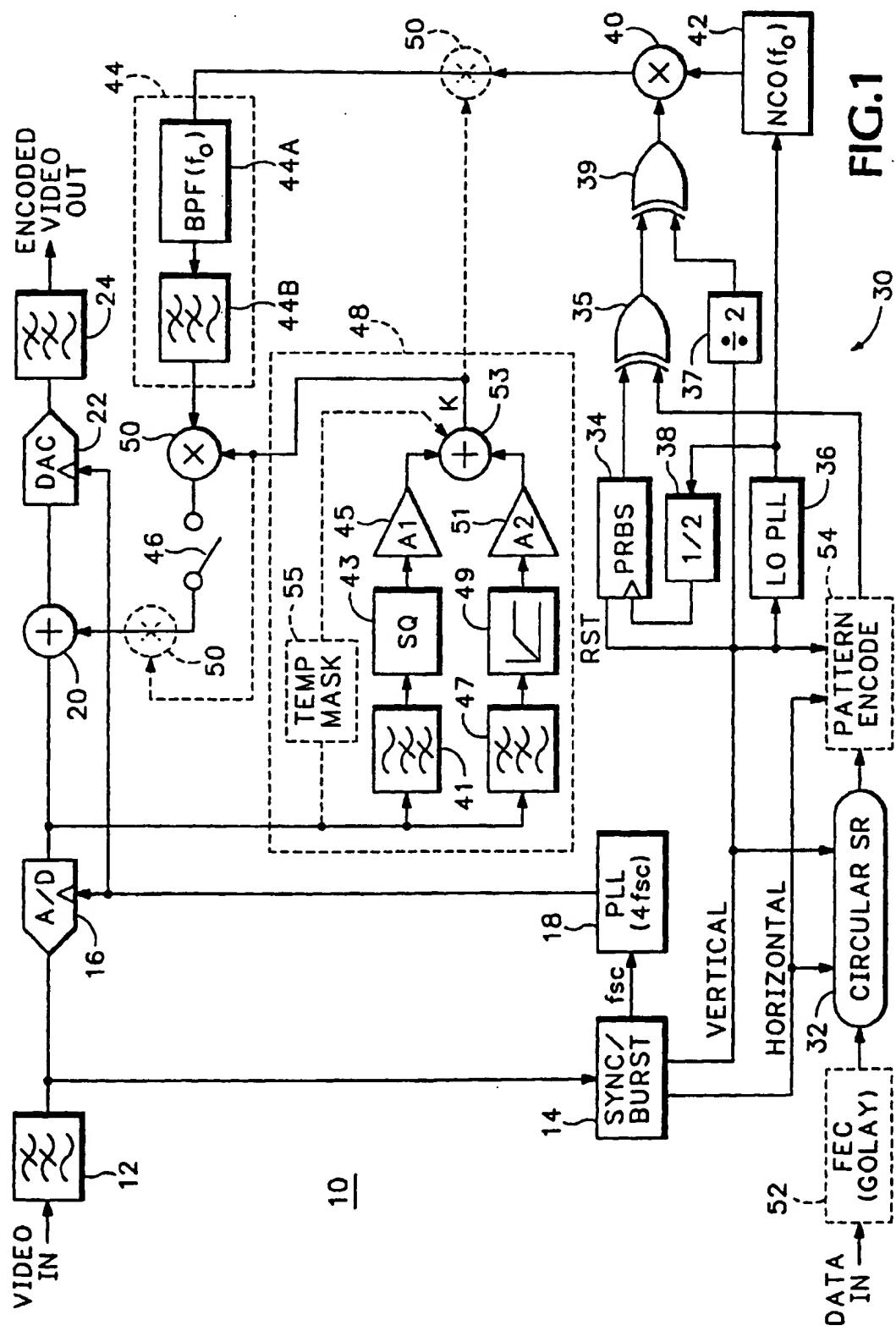
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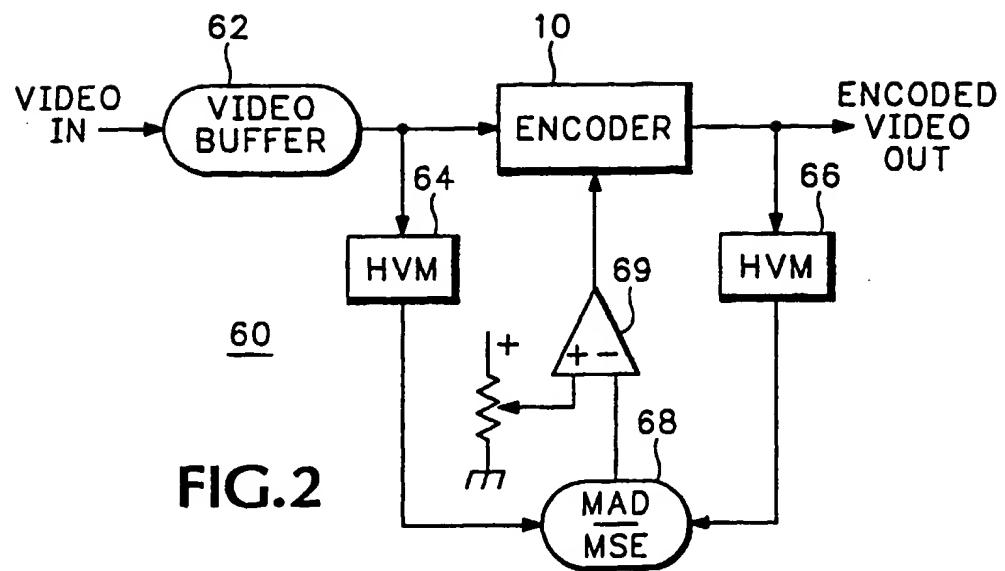
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FIG



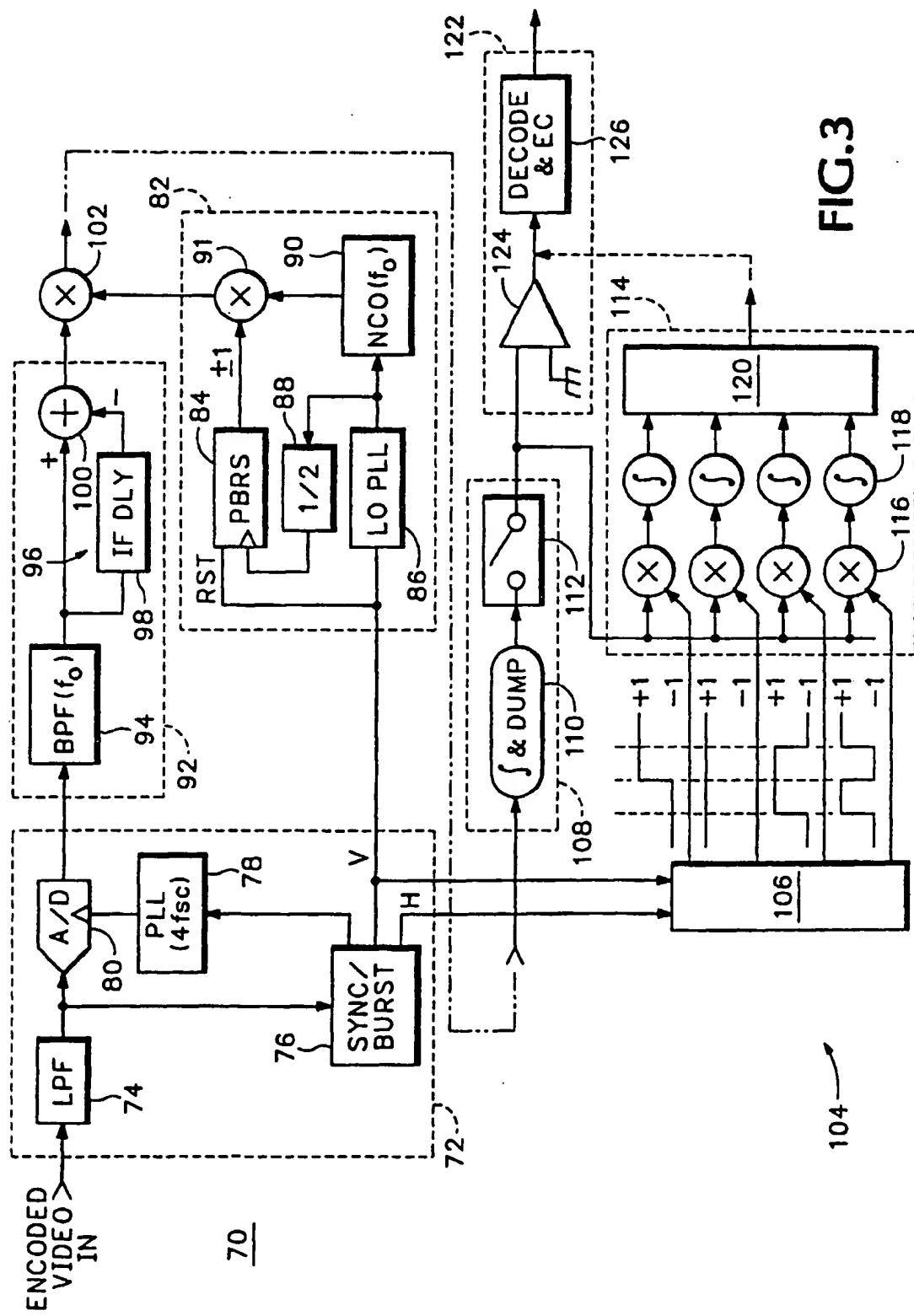


FIG. 3